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Robust Hybrid Kalman Filter for a Class of Nonlinear Systems

Bharani P. Malladi, Ricardo G. Sanfelice, and Eric A. Butcher

Abstract—Motivated by real-world applications with intermittent sensor data, an extended Kalman filter is formulated as a hybrid system and constructive conditions on its parameters guaranteeing an asymptotic stability property are provided. The dynamical properties of the estimation error are first characterized infinitesimally so to yield bounds on the rate of convergence and overshoot that depend on the parameters. By recasting the problem as the stabilization of a compact set, robustness properties of the proposed algorithm in the presence of disturbances in the system dynamics as well as measurement noise in the output are established. The proposed strategy is applied to spacecraft relative motion control with position-only measurements.

I. INTRODUCTION

The Kalman filter and its variants are among the most widely used estimation techniques in real-world applications in the presence of limited sensor data. Due to advances in the field of nonlinear observer design in recent times (see [1], [2] and the references therein), there have been a renewed interest in modeling Kalman filters as high gain observers. For this purpose, several strategies have been suggested for transforming an extended Kalman filter into an adaptive high-gain observer [3]. To keep the observer design more realistic, a continuous-discrete version of the high-gain extended Kalman filter for a control affine system with periodic measurements was explored in [4] and multi-rate sampling is presented in [5]. These observers are based on a canonical form of uniformly observable systems and present a solution-based approach to demonstrate the stability of the error dynamics under global Lipschitz assumptions. In addition, [5] presents the results on preservation of observability under multi-rate sampling. Most recently, in [6], a Luenberger-like version of these high-gain observers for interconnected systems was presented in the hybrid system framework [7]. In contrast to the above mentioned high-gain techniques in which, the correction gain associated with the new measurement is obtained by integrating a continuous-discrete time Riccati equation, the results in [8] and [9] consider a constant correction term with uniform and nonuniform sampled measurements respectively. Extending similar ideas to linear continuous-discrete time networked control systems, [10] presents observer design with various static and dynamic time-scheduling protocols. A trajectory-based stability analysis is presented using small-gain arguments. For state affine systems, [11] presents a continuous-discrete observer that establishes exponential stability property for the error dynamics with a regularly persistent input. A similar continuous-discrete observer with variable sampling period for measurements is presented in [12].

In this paper, we propose an extended Kalman filter (without high-gain) in its deterministic form with hybrid dynamics and present its associated stability analysis. More precisely, we propose an extended Kalman filter with nonuniform sampled measurements generating an estimate that changes discretely, when the information arrives, and also continuously in between such events. Particularly, in contrast to [5], where, a time varying output matrix formulation is used to represent nonuniform sampled measurements, we consider a constant output matrix. In addition, our results allow for the measurements of the output of the plant to be only available intermittently, thus only allowing the observer to update the estimates of the state at aperiodic time instances within a window. We present sufficient conditions involving the parameters defining that window to achieve convergence of the error to zero using a Lyapunov analysis for hybrid systems. Further, we also show that the proposed hybrid Kalman filter is robust to measurement noise in the outputs and perturbations associated with the system dynamics by presenting input-to-state stability analysis.

This work is motivated by the recent advances in applying hybrid system theory for formation flying of spacecraft and proximity operations (e.g. rendezvous and docking) [13]. In [14] and in [15] the authors have shown that considering the full nonlinear dynamics of the spacecraft will result in improved state estimation in otherwise weakly observable or unobservable linear models. Hence, we consider linearized spacecraft relative motion equations and apply our hybrid Kalman filter to estimate the state from aperiodic position measurements. In addition, for this application, the spacecraft is to be controlled via thrusters similar to the ideas presented in [13].

The remainder of this paper is organized as follows. The notation used in paper is in Section I-A. In Section II, we state the problem along with preliminaries on hybrid systems and propose an observer design. The hybrid model of the closed-loop system and its stability analysis are discussed in Section III. A robustness analysis for the closed-loop system is considered in Section III-D. Numerical results for the relative orbit estimation of a follower spacecraft relative to a target in a reference circular orbit using position-only measurements is in Section IV. Proofs of the results will be published elsewhere.

A. Notation

The following notation and definitions are used throughout the paper. \( \mathbb{R} \) denotes the real numbers. \( \mathbb{R}_n \) denotes \( n \)-dimensional Euclidean space. \( \mathbb{Z} \) denotes the integers. \( \mathbb{R}_{\geq 0} \)
denotes the nonnegative real numbers, i.e., $\mathbb{R}_{\geq 0} = [0, \infty)$. $\mathbb{N}$ denotes the natural numbers including 0, i.e., $\mathbb{N} = \{0, 1, \ldots\}$. $\Pi_{\geq 0}$ and $\Pi_{> 0}$ denote the set of positive semidefinite and positive definite, symmetric matrices, respectively. Given a set $A \subset \mathbb{R}^n$ and a point $x \in \mathbb{R}^n$, $|x|_A := \inf_{y \in A} |x-y|$. The equivalent notation $[x \ trans \ y]$ and $(x, y)$ is used for vectors, $x, y \in \mathbb{R}^n$. Given a vector $y \in \mathbb{R}^n$, $|y|$ denotes its Euclidean norm. For a generic vector norm $\| \cdot \|$, its corresponding induced matrix norm on $\mathbb{P}$ is given by $\|P\|$. Given a symmetric positive matrix $P$, $\lambda(P)$ denotes its eigenvalue. $I_s$ denotes the $s$-dimensional identity matrix. A function $\alpha : \mathbb{R}_{\geq 0} \rightarrow \mathbb{R}_{\geq 0}$ is said to belong to class-$\mathcal{K}$ if it is continuous, zero at zero, and strictly increasing. A function $\alpha : \mathbb{R}_{\geq 0} \rightarrow \mathbb{R}_{\geq 0}$ is said to belong to class-$\mathcal{K}_{\infty}$ if it belongs to class-$\mathcal{K}$ and is unbounded. A function $\beta : \mathbb{R}_{\geq 0} \times \mathbb{R}_{\geq 0} \rightarrow \mathbb{R}_{\geq 0}$ is said to belong to class-$\mathcal{KL}$ if it is nondecreasing in its first argument, nonincreasing in its second argument, and $\lim_{s \searrow 0} \beta(s, t) = \lim_{t \rightarrow \infty} \beta(s, t) = 0$.

II. PROBLEM STATEMENT

A. System design

In this paper, we consider control affine nonlinear systems of the form

$$
\dot{\eta} = f(\eta) + \sum_{i=1}^{n} \xi_i(\eta) u_i \quad \eta \in \mathbb{R}^n, u_i \in \mathbb{R}
$$

(1)

where $f : \mathbb{R}^n \rightarrow \mathbb{R}^n$, $\xi_i : \mathbb{R}^n \rightarrow \mathbb{R}^n$, $y \in \mathbb{R}^r$ is the measured output, and $h : \mathbb{R}^r \rightarrow \mathbb{R}^s$, that can be written in the normal form

$$
\begin{align*}
\dot{\eta}_1 &= \eta_2 + \sum_{i=1}^{n} \xi_{1,i}(\eta_1) u_i \\
\dot{\eta}_2 &= \eta_3 + \sum_{i=1}^{n} \xi_{2,i}(\eta_2, \eta_1) u_i \\
& \vdots \\
\dot{\eta}_{N-1} &= \eta_N + \sum_{i=1}^{n} \xi_{N-1,i}(\eta_1, \eta_2, \ldots, \eta_{N-1}) u_i \\
\dot{\eta}_N &= \varphi(\eta) + \sum_{i=1}^{n} \xi_{N,i}(\eta_1, \eta_2, \ldots, \eta_{N-1}, \eta_N) u_i \\
y &= \eta_1
\end{align*}
$$

(2)

on $\mathbb{R}^r$, where $\eta = (\eta_1, \eta_2, \ldots, \eta_N)$, $N \in \mathbb{N}_{\geq 0}$, $\eta_i \in \mathbb{R}^r$, and the functions $\varphi : \mathbb{R}^n \rightarrow \mathbb{R}^r$, $\xi_{i,j} : \mathbb{R}^n \rightarrow \mathbb{R}^r$, $i \in \{1, 2, \ldots, n\}$ are globally Lipschitz. Note that such a transformation and the properties of the associated functions can also be considered on an open subset $\Omega \subset \mathbb{R}^r$ if a normal form exists on it, but for simplicity we work on $\mathbb{R}^r$. System (2) can be compactly written as

$$
\begin{align*}
\dot{\eta} &= \eta_1 + b(\eta, u) \\
y &= H \eta
\end{align*}
$$

(3)

where $b(\eta, u)$ is lower block-triangular; $k$th component of $b_k$ depends only on $(\eta_1, \eta_2, \ldots, \eta_k)$ and the components $(\eta, u) \rightarrow b_k(\eta, u)$ are compactly supported with respect to all their arguments. The existence of the normal form (3) requires the nonlinear system (1) to have specific observability properties, including instantaneously uniform and strong differential observability; see [16], [11].

Let $t \rightarrow \eta(t)$ be a solution to the system (3) on $\mathbb{R}^r$ an input $u \in U$, where $U$ is a set of measurable functions with values in $\mathbb{U} \subset \mathbb{P}^r$ that are bounded on any subset of $[0, \infty)$. For each such $(\eta, u)$, we define $t \rightarrow A(t)$ as the linearization of (3). To this function, we associate the following family of linear time-varying systems with state $\zeta \in \mathbb{R}^n$ and input $\nu \in \mathbb{R}^p$:

$$
\dot{\zeta} = \overset{\sim}{A}(t) \zeta + \overset{\sim}{B}(t) \nu,
$$

(4)

where $\overset{\sim}{A}(t) := A + \frac{\partial b}{\partial \eta}(\eta(t), u(t))$, and $\overset{\sim}{B}(t) := \frac{\partial b}{\partial u}(\eta(t), u(t))$ for all $t \geq 0$. Below, $\Psi_0$ denotes the state transition matrix for (4) that satisfies

$$
\frac{d\Psi_0(t, s)}{dt} = A(t)\Psi_0(t, s), \quad \Psi_0(s, s) = I.
$$

(5)

for all $(t, s) \in [0, \infty) \times [0, \infty)$. Next, we consider the following assumptions.

Assumption 2.1: Given the normal form (3), there exists $\tau > 0$ such that for each $\eta \in \mathbb{R}^r$ there exists $\epsilon > 0$ such that

$$
\int_0^{\tau} \| \Psi_0(t, s) \|^2 H^T H \Psi_0(t, s) dt \geq \epsilon I,
$$

(6)

where $\Psi_0(t, s)$ for all $t, s \in [0, \tau]$ is given in (5).

Assumption 2.2: The input $u \in U$ is bounded and the function $(\eta, u) \rightarrow b(\eta, u)$ is globally Lipschitz uniformly in $u$. In particular, there exists $L \geq 0$ such that $|b(\eta_1, u) - b(\eta_2, u)| \leq L|\eta_1 - \eta_2|$ for any $\eta_1, \eta_2 \in \mathbb{R}^n$ and any $u \in U$.

Assumption 2.1 requires uniform infinitesimal observability (or uniform reconstructibility) of the normal form (3). This property holds for free (and actually uniformly on $c$) for systems in such normal form with a bilinear nonlinearity $b$; see [1, Lemma 2.11]. Essentially, this assumption guarantees a lower and upper bound on the information matrix of the Kalman filter; see [17, Proposition 2.2 and Proposition 2.4].

For the systems satisfying the assumptions above, we propose a Kalman-like hybrid observer that guarantees an explicit $\mathcal{KL}$ bound on the estimation error in terms of the parameters for all (hybrid) time instances in the domain of definition of the trajectories.

B. Well-posed hybrid systems

Hybrid systems are dynamical systems with both continuous and discrete dynamics. In this paper, we consider the framework for hybrid systems outlined in [7], [18], where a hybrid system $\mathcal{H} = (C, f, D, g)$ is defined by the following objects: a map $f : \mathbb{R}^n \rightarrow \mathbb{R}^n$ called the flow map; a map $g : \mathbb{R}^n \rightarrow \mathbb{R}^n$ called the jump map; a set $C \subset \mathbb{R}^n$ called the flow set; and a set $D \subset \mathbb{R}^n$ called the jump set. The flow map $f$ defines the continuous dynamics on the flow set $C$, while the jump map $g$ defines the discrete dynamics on the jump set $D$. These objects are referred to as the data of the hybrid system $\mathcal{H}$. Given a state $\chi$ of the hybrid system $\mathcal{H}$, the notation $\chi^+$ indicates the values of the state after the
jump. A solution $\phi$ to $\mathcal{H}$ is given on extended time domain, called hybrid time domain, that is parametrized by the pairs $(t, j)$, where $t$ is the ordinary time component and $j$ is a discrete parameter that keeps track of the number of jumps; see [7, Definition 2.6]. Given a solution $\phi$ to $\mathcal{H}$, the notation $\text{dom } \phi$ represents its domain, which is a hybrid time domain. A solution to $\mathcal{H}$ is said to be nontrivial if $\text{dom } \phi$ contains at least one point different from $(0, 0)$, complete if $\text{dom } \phi$ is unbounded, and maximal if it cannot be extended, i.e., it is not a truncated version of another solution. The set $S_\mathcal{H}(\xi)$ denotes the set of all maximal solutions to $\mathcal{H}$ from $\xi$. This framework also permits explicit modeling of perturbations in the system dynamics, a feature that is very useful for robust stability analysis of dynamical systems; see [7] for more details.

C. Observer design

Given system (3), the measurements of $y = h(\eta)$ are available intermittently at time instances $t_k$, $k \in \mathbb{N}$ that are separated by $\tau \in [T_{\text{min}}, T_{\text{max}}]$ seconds, where $0 < T_{\text{min}} \leq T_{\text{max}}$. The scalars $T_{\text{min}}$ and $T_{\text{max}}$ define the lower and upper bounds, respectively, for the time in between the measurement events. These dynamics implement the following mechanism: whenever the timer $\tau$ reaches an arbitrary value in the closed set $[T_{\text{min}}, T_{\text{max}}]$, a new measurement is available and the timer is reset back to zero from where it counts towards $[T_{\text{min}}, T_{\text{max}}]$ again. The dynamics for the timer $\tau$ are given by the hybrid model

$$
\begin{align*}
\tau^- &= 1, \quad \tau \in [0, T_{\text{max}}], \\
\tau^+ &= 0, \quad \tau \in [T_{\text{min}}, T_{\text{max}}].
\end{align*}
$$

(7)

With the availability of a new measurement $y \in \mathbb{R}^r$, motivated by the results in [1], [3], [19], we propose the following hybrid observer with state $x_c := (\eta_c, S, \tau) \in \mathbb{R}^n \times \Pi_{\mathbb{R}^n \times [0, T_{\text{max}}]} := X_c$

$$
\begin{align*}
\dot{x}_c &= f_c(x_c, u) \quad (x_c, u) \in C_c \times U, \\
\chi_c^+ &= g_c(x_c), \quad \chi_c \in D_c,
\end{align*}
$$

(8)

where the maps $f_c : X_c \times U \rightarrow X_c$, $g_c : X_c \rightarrow X_c$ and the sets $C_c \subset X_c$, $D_c \subset X_c$ are

$$
\begin{align*}
f_c(x_c, u) &:= \begin{bmatrix} A_{\eta} + b(\eta_c, u) \\ -A^T S - SA - SQS \end{bmatrix} \forall (x_c, u) \in C_c \times U, \\
g_c(x_c) &:= \begin{bmatrix} \eta_c + (\mathbb{S})^{-1} H^T (R)^{-1} (y - h(\eta_c)) \\ 0 \end{bmatrix} \forall x_c \in D_c,
\end{align*}
$$

$$
C_c := \mathbb{R}^n \times \Pi_{0 \geq 0 \times [0, T_{\text{max}}]}, \\
D_c := \mathbb{R}^n \times \Pi_{0 \geq 0 \times [T_{\text{min}}, T_{\text{max}}]};
$$

with a slight abuse of notation $A := A + \frac{\partial h}{\partial \eta}(\eta_c, u)$, as defined below (4), $\mathbb{S} := S + H^T (R)^{-1} H$; the $Q \in \mathbb{R}^n$ and $R \in \mathbb{R}^n$ are similar to the covariance matrices of the state noise and output noise in the stochastic context, respectively.

The components of the observer state $x_c$ consist of the estimated system state $\eta_c$, the error information matrix $S$, and a timer $\tau$ that triggers measurement errors at isolated time instances $t_k$, $k \in \mathbb{N}$. In addition, the input $u \in U$ is available to the observer, i.e., the input is not sampled.

Note that the hybrid observer dynamics given in (8) resemble the continuous-discrete extended Kalman filter presented in [1, Section 2.4.4] but without high gain. In particular, the flow dynamics $(f_c, C_c)$ resemble the ‘prediction step’ given in [1, equation (104)] and the jump dynamics $(g_c, D_c)$ resemble the ‘innovation step’ [1, equation (105)]. In addition to adapting the extended Kalman filter formulation into a hybrid system framework, the observer proposed in (8) incorporates a time window $[T_{\text{min}}, T_{\text{max}}]$, as opposed to the periodic measurement updates in [1] and time varying output matrix to represent aperiodic measurement updates in [5]. Such mechanism is explicitly modeled as part of the dynamics of the hybrid system. In [12], a continuous-discrete observer for a state affine system is designed based on variable sampling time that is used as a tuning parameter to compute measurement time online resembling an event-triggered methodology. In contrast, the observer proposed in (8) accommodates sporadic sensor measurements for uniformly observable systems.

III. STABILITY PROPERTIES OF THE HYBRID CLOSED-LOOP SYSTEM

To estimate the state $\eta \in \mathbb{R}^n$ from the intermittent measurements $y \in \mathbb{R}^r$, we consider the plant model in (3) interconnected with the observer given in (8). The resulting hybrid closed-loop system $\mathcal{H} := (C, f, D, g)$ has state $\chi := (\eta, \eta_c, S, \tau) \in \mathbb{R}^n \times X_c := X$ and dynamics given by

$$
\begin{align*}
\dot{\chi} &= f(\chi, u) \quad (\chi, u) \in C, \\
\chi^+ &= g(\chi) \quad \chi \in D,
\end{align*}
$$

(9)

where $f : X \times U \rightarrow X$, $g : X \rightarrow X$ and the sets $C \subset X$, $D \subset X$ are

$$
\begin{align*}
f(\chi, u) &:= \begin{bmatrix} A_{\eta} + b(\eta, u) \\ -A^T S - SA - SQS \end{bmatrix} \forall (\chi, u) \in C \times U, \\
g(\chi) &:= \begin{bmatrix} \eta_c + (\mathbb{S})^{-1} H^T (R)^{-1} (y - h(\eta_c)) \\ 0 \end{bmatrix} \forall \chi \in D.
\end{align*}
$$

$$
C := \mathbb{R}^n \times C_c, \\
D := \mathbb{R}^n \times D_c.
$$

With the construction above, the hybrid closed-loop system $\mathcal{H}$ satisfies the following property.

Lemma 3.1: The hybrid system $\mathcal{H}$ satisfies the hybrid basic conditions, i.e.,

(A1) $C$ and $D$ are closed subsets of $X$.

(A2) $f : X \times U \rightarrow X$ is continuous.

(A3) $g : X \rightarrow X$ is continuous.

Next, we consider the following property on the timer $\tau$.

Lemma 3.2: Given a solution $\phi$ to $\mathcal{H}$, let $t_j$ be defined such that a jump occurs at $(t_j, j - 1)$ with $j \in \mathbb{N} \setminus \{0\}$. Then,

$$
\begin{align*}
t_j &\leq 1 + \frac{t}{T_{\text{min}}},
\end{align*}
$$

(10)

for all $(t, j) \in \text{dom } \phi$, $j \geq 1$.

A. Error Dynamics

The set of interest for closed-loop hybrid system $\mathcal{H}$ in (9) is the closed set

$$
\mathcal{A} := \{ \chi \in X : \epsilon = 0, \tau \in [0, T_{\text{max}}] \},
$$

(11)
where \( e := \eta - \eta_c \). The dynamics of \( e \in \mathbb{R}^n \) from (9) are hybrid and given by

\[
\dot{e} = \hat{\eta} - \eta_c = Ae + b(\eta, u) - b(\eta_c, u) \quad (\chi, u) \in C \times U, \\
e^+ = e - (\ast)^{-1}H^T(\frac{\tau}{\sigma})^{-1}He \quad \chi \in D,
\]

(12)

here \( C := \mathbb{R}^n \times C_c, D := \mathbb{R}^n \times D_c \), the flows occur when the timer \( \tau \in [0, T_{\max}] \) and jumps occur when \( \tau \in [T_{\min}, T_{\max}] \).

First, we provide a matrix version of [1, Lemma 2.21] which will be used to derive the results in this paper.

Lemma 3.3: If \( F = S^{-1} \) is a symmetric positive definite matrix, \( R \in \mathbb{R}^n \) and \( \tau > 0 \), then

\[
(P + PH^T(\frac{\tau}{\sigma})^{-1}HP)^{-1} = P^{-1} - H^T(HPH^T + \frac{\tau}{\sigma})^{-1}H.
\]

In addition, the following infinitesimal properties of the function \( V : \chi \mapsto \mathbb{R}_{\geq 0} \), will be used for the main result presented in Section III-C.

Proposition 3.4: Given the hybrid error dynamics \( e \in \mathbb{R}^n \) in (12), and the dynamics of \( S \) in (9) with \( S \in \Pi \), the function

\[
V(\chi) := e^TSe
\]

has the following infinitesimal properties:

- for all \( \chi \in C \setminus A, u \in U \),
  \[
  \langle \nabla V(\chi), f(\chi, u) \rangle = e^T(-S \Phi)Se + 2e^T S(b(\eta, u) - b(\eta_c, u) - b^*(\eta_c, u)e),
  \]
  where \( b^* = \frac{\partial h}{\partial \eta_c} \).
- for all \( \chi \in D \setminus A \),
  \[
  V(g(\chi)) - V(\chi) = e^T(H^T(\sigma_{\min}^{-1}H^T + R/\tau)^{-1}He).
  \]

B. Analysis of \( S \) Component of Solutions

To establish a desired stability result for the hybrid closed-loop system (9), we perform the following analysis on the \( S \) component of its solution \( \phi := (\phi_\eta, \phi_{\eta_c}, \phi_S, \phi_\tau) \).

Assumption 3.5: The constants \( T_{\min}, T_{\max} \), the matrix \( \bar{A} := A + \frac{\partial h}{\partial \eta_c} \), and the positive definite symmetric matrices \( Q, \bar{R} \) satisfy the following conditions: and there exists positive constants \( a, q, \bar{q}, q, \bar{q}, \tau \) such that

1) \( |\bar{A}| \leq a \quad \forall (\eta_c, u) \in \mathbb{R}^n \times U \\
2) 0 < q_1 \leq Q \leq q_1 \\
3) 0 < \bar{q}_1 \leq \bar{R} \leq \bar{q}_1 \\
4) 0 < T_{\min} \leq T_{\max}

Lemma 3.6: Suppose Assumption 3.5 holds, and that a positive real number \( 0 < k_1 < k_2 := 4a/q \) and a compact set

\[
K \subset \{ S \in \Pi_{\geq 0} : k_1 I \leq S \leq k_2 I \}
\]

are given. Let the positive real number \( \tau \) satisfy the condition (6) in Assumption 2.1. Let \( \phi_S \) be the \( S \) component of the maximal solution \( \phi \) to the hybrid system \( \mathcal{H} \) of \( \phi(0,0) \in \mathbb{R}^n \times \mathbb{R}^n \times K \times [0, T_{\max}] \) to the closed-loop system in (9) is bounded and does not blow up in finite time. Also, \( g(D) \subset C \cup D \) which shows that the solution \( \phi \) to system \( \mathcal{H} \) do not jump out of \( C \cup D \). Therefore, since conditions (b) and (c) in [7, Proposition 6.10] are not satisfied, we conclude that every maximal solution to the closed-loop system \( \mathcal{H} \) is complete.

D. Input-to-state stability

To take our stability analysis close to the real world problems, let us consider that the plant (9) is affected by unmodeled dynamics given by \( d_1 \in \mathbb{R}^{n \times 1} \), actuator error \( d_2 \in \mathbb{R}^{p \times 1} \) as following

\[
\dot{\eta} = f(\eta, u + d_2) + d_1.
\]

We also consider that the measurement noise \( d_3 \in \mathbb{R}^{r \times 1} \) is added to the output, \( y = h(\eta) + d_3 \).

Then, denoting by \( d \) the signals \( d_i, i \in \{1,2,3\} \) extended to the state space of \( \chi := (\eta, h, S, \tau) \in X \), the closed loop system \( \mathcal{H} \) in (9) results in the perturbed closed-loop system \( \mathcal{H}_d \) with dynamics

\[
\dot{\chi} = f_d(\chi, u + d_2) + d_p(\chi, u + d_2) + d_p \in C_d \times U, \\
\chi^+ = g_d(\chi, y + d_3) \quad (\chi, y + d_3) \in D_d \times \mathbb{R},
\]

(18)
here \(d_p := \begin{bmatrix} d_1 & 0_{nx \times 1} & 0_{nx \times n} \end{bmatrix}^T\), the maps \(f_d : X \times U \to X\), \(g_d : X \times \mathbb{R}^r \to X\) and the sets \(C_d \subset X\), \(D_d \subset X\), respectively. Next, we show that the semiglobal-asymptotic stability of the compact set \(A\) in (11) for the hybrid system \(\mathcal{H}_d\) in (18) is robust to perturbations and is input-to-state stable (ISS), namely

**Definition 3.9:** A hybrid system \(\mathcal{H}_d\) with input \(d\) is input-to-state stable with respect to \(A\) if there exists \(\beta \in \mathcal{KL}\) and \(\kappa \in K\) such that each solution pair \((\phi, d)\) to \(\mathcal{H}\) satisfies

\[
\|\phi(t, j)\|_A \leq \max \{\beta(\|\phi(0, 0)\|_A, t + j), \kappa(\|d(t, j)\|)\}.
\]

To establish ISS result, let us first consider the following assumption on the input with perturbations.

**Assumption 3.10:** The input \(u \in U\) and input perturbation \(t \mapsto d_2(t) \in \mathbb{R}^p\) are bounded and the function \((\eta, u + d_2) \mapsto b(\eta, u + d_2)\) is globally Lipschitz uniformly in \(u + d_2\). In particular, there exists \(L \geq 0\) such that \(\|b(\eta_1, u + d_2) - b(\eta_2, u + d_2)\| \leq L(\|\eta_1 - \eta_2\|)\) for any \(\eta_1, \eta_2 \in \mathbb{R}^n\).

**Theorem 3.11:** Let the input \(u \in U\) and the function \(b\) satisfy Assumption 3.10 with \(L > 0\). Suppose that the constants \(T_{\text{min}}, T_{\text{max}}\) and that there exists positive constants \(a, b, c, \sigma, \tau\), such that the matrix \(A = A + \frac{\sigma}{\rho_k} (\eta, u)\) and the positive definite symmetric matrices \(Q, R\) satisfy the conditions in Assumption 3.5. Let the positive real numbers \(\tau\) satisfy the condition (6) in Assumption 2.1. Given positive real numbers \(0 < k_1 < k_2 := 4a/q\) and a compact set \(K \subset \{S \in \Pi_{\geq 0} : k_1 I \leq S \leq k_2 I\}\), if \(k_1 \approx 1, k_2 \approx 2a\gamma\) and \(k_3\) are such that (15), and (16) hold, where \(\kappa\) is any positive number sufficiently small to verify (16), then the hybrid system \(\mathcal{H}_d\) with input \(u\) is ISS with respect to the set \(A\), namely, for each solution pair \((\phi, d)\) with \(\phi \in \mathcal{S}_d(\phi(0, 0))\), where \(\phi(0, 0) \in \mathbb{R}^n \times \mathbb{R}^n \times K \times [0, T_{\text{max}}]\) satisfies (19), and \(\beta\) is a class-\(\mathcal{KL}\) function given by

\[
\beta(r, s) := \alpha_1^{-1}(\alpha_2(r) \exp(m - \gamma(s))) \quad \forall r, s \geq 0,
\]

\[m := \sigma T_{\text{max}} + \ln 2, \quad \rho := (qk - 4k^2) + \sigma \text{ and for each} \quad s \geq 0, \quad \gamma(s) := (\rho - \sigma T_{\text{max}}) s, \quad \alpha_1, \alpha_2/s\text{ are class-}K_{\infty}\text{ functions, the class-}K\text{ function} \quad \kappa(r) := \alpha_1^{-1}(\alpha_2(\max\{a_1, a_2, a_3\})r) \quad \forall r \geq 0,
\]

that holds for input \(d\) at \((t, j)\) as

\[
\begin{aligned}
\alpha_1 & := \sqrt{\frac{\exp(\sigma T_{\text{max}}) (q \gamma^2 + 4k^2)}{4kL}}, \\
\alpha_2 & := \frac{2 \exp(\sigma T_{\text{max}} - \frac{1}{k_2})}{\rho_k}, \\
\alpha_3 & := \sqrt{\frac{\exp(\sigma T_{\text{max}} - \frac{1}{k_2})}{\rho_k L}}.
\end{aligned}
\]

IV. APPLICATION

A. Spacecraft Nonlinear Model

We consider a nonlinear model of relative circular motion of a chaser spacecraft relative to a target spacecraft resolved in the local vertical local horizontal frame (denoted by \(H\)) (see [20], Chapter 14) for more details given by

\[
\frac{d^2}{dt^2}(\phi^H) = -2\omega^H \times \phi^H - \phi^H \times (\omega^H \times \phi^H) + \mu \omega^H \times \phi^H + \eta^H,
\]

where \(\phi^H := (r, \dot{r}, \omega, \theta, \psi) \in \mathbb{R}^3\) and \(\omega^H := (x, y, z) \in \mathbb{R}^3\) and \(\phi^H := (\dot{x}, \dot{y}, \dot{z}) \in \mathbb{R}^3\) are position and velocity, respectively, \(r_0\) is the orbit radius of the target spacecraft, \(\phi^H := \frac{1}{m_0} (F_x F_y F_z) \in \mathbb{R}^3\) is the input with \(F_x, F_y, F_z\), the control forces in the \(x, y\) and \(z\) directions, \(m_0\) is the mass of the chaser, \(\omega^H := (0, 0, n)^T\), \(n := \sqrt{\mu / r^3}\) where \(\mu\) is the gravitational parameter of the Earth, \(\eta^H := (r + x, y, z) \in \mathbb{R}^3\) and \(\|\eta^H\| = \sqrt{(r + x)^2 + y^2 + z^2}\). In addition, the input \(\eta^H\) has a maximum thrust constraint \(\|\eta^H\| \leq u_{\text{max}}, u_{\text{max}} > 0\).

1) Position only measurements: With position only measurements, the linearized spacecraft equations in (20) are in the desired normal form given by

\[
\begin{aligned}
\dot{\eta} &= A\eta + b(\eta, u) \\
\eta &= C\eta
\end{aligned}
\]

where the state vector \(\eta := (\phi^H, \hat{\phi}^H) \in \mathbb{R}^6\). The block-antisymmetry matrix and the output matrices are

\[
A := \begin{bmatrix} 0_{3 \times 3} & I_{3 \times 3} \\ 0_{3 \times 3} & 0_{3 \times 3} \end{bmatrix}, \quad C := \begin{bmatrix} I_{3 \times 3} \\ 0_{3 \times 3} \end{bmatrix},
\]

respectively. The lower block-triangular function \(b(\eta, u) := A_d\eta + \frac{1}{m_0} B u,\)

The system dynamics in (21) is linear, observable and in the normal form (3) thus satisfying Assumption 2.1 and with maximum thrust constraint \(\|\eta^H\| \leq u_{\text{max}}, u_{\text{max}} > 0\), satisfies the Assumption 2.2, respectively. Note that that the linearized equation of motion (21) represent the so-called Clohessy-Wiltshire-Hill equations. Since \(r_o < r_t\), for all \(\eta, \eta_t \in \mathbb{R}^6\), the lower block triangular terms of the spacecraft linearized equations are bounded by \(b(\eta, u) - b(\eta, u) \leq L|\eta - \eta_t|\), where \(L \approx 2n = 0.0021\). To estimate the state \(\eta\) from position \((\phi^H)\) only measurements, we implement a hybrid observer presented in (9) on the resulting plant. Next, we design a feedback controller, to which the state estimates \(\eta_t\) is fed similar to the controller design presented in (13).

2) Simulations: To simulate the spacecraft dynamics, we use \(n = \sqrt{\mu / r^3}, \quad \mu = 3.986 \times 10^{14} m^3/sec^2, \quad r_0 = 71000000\), \(m_0 = 500 K_g\) in these simulations. In these simulations, the chaser starts at a distance of no more than \(r_{\text{max}} = 10 K m\) away from the target. In addition, the thrusters have a maximum input of \(u_{\text{max}} = 0.02 m/sec^2\). As we consider full nonlinear relative motion equations for our application, a process noise (zero-mean Gaussian noise) with variance \((10^{-3})^2\) is added to the system dynamics to compensate for any additional unmodeled dynamics. Next, to simulate position measurements, a zero-mean Gaussian noise with variance \((25 m)^2\) is considered.

With these mission parameters, simulations for the entire closed-loop system are performed for the chaser starting from \(\eta\) corresponding to various initial conditions in the 10Km radius with a initial velocity \(\rho(\dot{x}, 0), y(0), \dot{z}(0)) \in [0, 0.8606 m/sec]\). The initial condition for the estimated state is \(\eta_t(0, 0) \in [1000m 1000m 0m/\text{sec} 0m/\text{sec}]\). For the controller design, an LQR controller is implemented with appropriate parameters satisfying input constraint \(\|u\|_\infty \leq 0.02 m/sec^2\).

With the choice of \(T_{\text{min}} = 2\text{sec}\), the parameter \(\sigma\) is chosen sufficiently small to verify equation (16) and \(T_{\text{max}} <\)
$T_{\text{min}} (1 + 2) \leq 6 \text{sec.}$ Following this analysis the measurements in the simulation are updated intermittently at time instances $t_k, k \in \mathbb{N}$ that are separated by $\tau \in [2 \text{sec}, 6 \text{sec}]$.

Figure 1 shows the induced norm of the information matrix $\mathbf{S}$ satisfies $0 < h_1 < h_1 < h_2 := 40/\gamma$ for a simulation time of $12 \times 10^3$ seconds. The trajectories of the chaser spacecraft for various initial conditions are shown in Figure 2, while the error, timer dynamics are presented in Figure 3. This shows that the hybrid Kalman filter works effectively to reduce the differences between the true and computed measurements for the estimated orbit in the presence of bounded inputs.

V. CONCLUSION

In this paper, a robust hybrid Kalman filter (in its deterministic form) with intermittent measurement data is presented. The problem of stabilizing the hybrid closed-loop system to a desired set was solved with assumptions on a certain normal form. Note that such an assumption can be relaxed by using high-gain formulation. Additionally, it was established that the hybrid system is robust to perturbations. We also quantified an upper bound on the perturbations and stabilized using high-gain formulation. Additionally, it was established normal form. Note that such an assumption can be relaxed by to a desired set was solved with assumptions on a certain

\begin{thebibliography}{10}


Fig. 1. Bounds on the information matrix $\phi_2(t,j) \in K$

Fig. 2. Chaser true and estimated dynamics in XY and YZ, respectively

Fig. 3. Position estimate error (blue) and $5 - \sigma$ covariance matrix ($P$) bound (red) corresponding to the position-only estimation scenarios using full nonlinear dynamic models and timer dynamics with $\tau \in [2 \text{sec}, 6 \text{sec}]$